

F2E5216/TS1002 Adaptive Filtering and Change Detection

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Lecture 1

The goal of the course is to get an understanding for the theory of model based filtering and change detection, with particular attention to applications.

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The Books

Steven Kay: Selected parts of Fundamentals of statistical signal processing: Volume II, detection theory. Prentice Hall, 1998.

Only Lecture 1

Fredrik Gustafsson: Adaptive filtering and change detection. John Wiley & Sons, 2001.

Course Organization

- **Lectures and compendium:** Theory, Algorithms, Applications, Evaluation
- **Toolbox and manual:** Algorithms, Evaluation ↔ Implementations
- **Projects:** Application ↔ Evaluation of Algorithms
- **Exercises:** To each part available from homepage
- **Examination, 6 credits:** Three-day take-home exam.

Lecture plan

Lecture 1: Detection theory from Kay's book.

Lecture 2: Applications. Detection of mean changes: likelihood principles; Chapters 1, 2, 3 .

Lecture 3: Detection of mean changes: The parallel filter concept. Chapters 3,4, Appendix 4.

Lecture 4: Parameter estimation: recursive least squares. Residual based detection. Chapter 5

Lecture 5: Parameter estimation: parallel filters and filter bank change detection . Chapter 6-7

Lecture 6: State estimation: Kalman, EKF and particle filters. Residual based detection. Chapter 8

Lecture 7: State estimation: parallel filters and filter bank change detection. Chapters 9, 10. Given by Fredrik

Lecture 8: State estimation: the parity space approach. Chapter 11

Lecture 9: State estimation: parity space diagnosability and PCA.

Probability Theory

Lecture 1

- Basic Probability Theory
- Basic Detection Theory
- Neyman -Pearson's theorem
- Basic Hypothesis tests
- Generalized Likelihood Ratio Tests

Pages 1- 245 in Kay

Stochastic variable: X

Probability density function (PDF): $p(x)$

$$\text{Mean: } E(X) = \int_{-\infty}^{\infty} xp(x)dx$$

$$\text{Variance: } Var(X) = \int_{-\infty}^{\infty} (x - E(X))^2 p(x)dx$$

$$\text{Cummulative distribution function: } \Phi(x) = P(X < x) = \int_{-\infty}^x p(x)dx$$

$$\text{Right tail probability: } Q(x) = P(X > x) = \int_x^{\infty} p(x)dx$$

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Gaussian or Normal

$$N(\mu, \sigma^2)$$

$p(x)$	$\frac{1}{\sqrt{2\pi}\sigma^2} e^{-\frac{(x-\mu)^2}{2\sigma^2}}$
$E(x)$	μ
$Var(x)$	σ^2
$Q(x)$	$\approx \frac{1}{\sqrt{2\pi}(x-\mu)/\sigma} e^{-\frac{(x-\mu)^2}{2\sigma^2}}$
$E(x^n)$	$1 \cdot 3 \cdot 5 \dots (n-1)\sigma^2$
Matlab	<code>Q(x)=0.5*erfc((x-mu)/sigma/sqrt(2))</code>

Exponential

$$\text{Exp}(\mu)$$

$p(x)$	$\frac{1}{\mu} e^{-\frac{x}{\mu}}, x \geq 0$
$E(x)$	μ
$Var(x)$	μ^2
$Q(x)$	$e^{-\frac{x}{\mu}}$

Chi-squared with ν degrees

$$\chi_{\nu}^2$$

$p(x)$	$\frac{1}{2^{\nu/2}\Gamma(\nu/2)}x^{-\nu/2-1}e^{-x/2}, x > 0$
$\Gamma(u)$	$\int_0^\infty t^{u-1}e^{-t}dt, \Gamma(n) = (n-1)!$
$E(x)$	$\Gamma(u) = (u-1)\Gamma(u-1), \Gamma(0.5) = \sqrt{\pi}$
$V(x)$	ν
$Q(x)$	2ν
Comments	$e^{-x/2} \sum_{k=0}^{\nu/2-1} \frac{(x/2)^k}{k!}, \nu \geq 2$ and even $\chi_2^2 = Exp(2).$ If $x = \sum_{i=1}^{\nu} x_i^2, x_i \in N(0, 1)$, then $x \in \chi_{\nu}^2$ approximately Gaussian for large ν

Non-central chi-squared with ν degrees

$$\chi_{\nu}^2(\lambda)$$

$p(x)$	$\frac{1}{2} \left(\frac{x}{\lambda}\right)^{\frac{\nu-2}{4}} e^{-(x+\lambda)/2} I_{\nu/2-1}(\sqrt{\lambda x}), x > 0$
$I_r(u)$	$\sum_{k=0}^{\infty} \frac{(u/2)^{2k+r}}{k!\Gamma(r+k+1)}$ Bessel function
$E(x)$	$\nu + \lambda$
$V(x)$	$2\nu + 4\lambda$
Comments	If $x = \sum_{i=1}^{\nu} x_i^2, x_i \in N(\mu_i, 1)$, then $x \in \chi_{\nu}^2(\sum_i \mu_i)$

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Detection Principles

Given: Data $x[n]$ with known distribution under the *null hypothesis* H_0 and the *alternative hypothesis* H_1 , respectively.

General test: Form a test statistic $T(x)$.

Decide H_0 if $T(x) < \gamma$ (*threshold*)
Decide H_1 if $T(x) > \gamma$

Basic design parameters:

- *Probability of false alarm* (Type I error)
 $P_{FA} = P(H_1|H_0) = \int_{x:T(x)>\gamma} p(x|H_0)dx$
- *Probability of detection* (Power)
 $P_D = P(H_1|H_1) = \int_{x:T(x)>\gamma} p(x|H_1)dx$
- Miss $P_M = P(H_0|H_1) = 1 - P_D$ (Type II error)

The only 'hard' result in detection theory

Neyman-Pearson's theorem: To maximize the power P_D for a given probability of false alarm $P_{FA} = \alpha$, take

$$T(x) = \frac{p(x|H_1)}{p(x|H_0)}.$$

The test is called **LRT**, the **Likelihood Ratio Test**.

The threshold is found from

$$P_{FA} = \int_{x:T(x)>\gamma} p(x|H_0)dx = \alpha$$

Proof:

Define the LRT test $\mathcal{T} = \begin{cases} 1, & \text{if, } p(x|H_1) > \gamma p(x|H_0) \\ 0, & \text{if, } p(x|H_1) < \gamma p(x|H_0) \end{cases}$

Let $0 \leq \mathcal{T}' \leq 1$ denote any test with less or equal P'_{FA} .

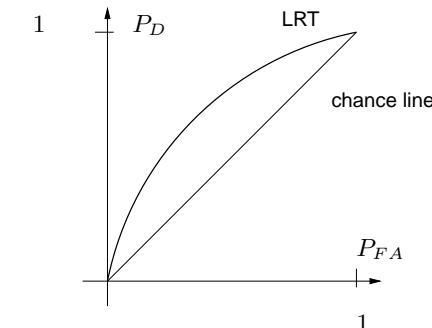
Key inequality (Check the signs!)

$$\int_{-\infty}^{\infty} (\mathcal{T} - \mathcal{T}')(p(x|H_1) - \gamma p(x|H_0))dx \geq 0$$

Re-arrange the integral: $(P_D - P'_D) \geq \gamma(P_{FA} - P'_{FA}) \geq 0$, where the last equality follows from the false alarm assumption.

This shows \mathcal{T} optimize the power! One has to be a bit more careful if $p(x|H_1) = \gamma p(x|H_0)$ has a non-zero probability

Receiver Operating Characteristics



The slope of the LRT ROC curve is γ !

The curve is concave.

The LRT maximizes any performance criterion that values low false alarm probabilities and/or high detection probabilities.

Proof of Slope

$$\begin{aligned} E(T(x)^n | H_1) &= \int \left[\frac{p(x|H_1)}{p(x|H_0)} \right]^n p(x|H_1) dx = \\ &\int \left[\frac{p(x|H_1)}{p(x|H_0)} \right]^{n+1} p(x|H_0) dx = E(T(x)^{n+1} | H_0) \end{aligned}$$

which can be written $\int t^n g(t|H_1) dt = \int t^{n+1} g(t|H_0) dt, \forall n$

where $g(t|H_i)$ is the PDF for the likelihood ratio under H_i . This means that $g(t|H_1) = t g(t|H_0)$. Now

$$P_D = \int_{\gamma}^{\infty} g(t|H_1) dt, \quad P_{FA} = \int_{\gamma}^{\infty} g(t|H_0) dt$$

and

$$\frac{dP_D}{dP_{FA}} = \frac{dP_D}{d\gamma} / \frac{dP_{FA}}{d\gamma} = \frac{g(t|H_1)}{g(t|H_0)} = \gamma$$

Observe that $\gamma \geq 0$

Sufficient Statistics

Replace x with the simpler sufficient statistics t ,

$$T(x) = \frac{p(x|H_1)}{p(x|H_0)} = \frac{g(t|H_1)}{g(t|H_0)}$$

Example: $x = (x_0 \dots x_{N-1})$, $x_i \in N(m, \sigma^2)$, take

$$t = \frac{1}{N} \sum_{i=0}^{N-1} x_i$$

Basic Hypothesis Test

Is there a known mean A in the observed signal?

$$H_0 : x[n] = w[n]$$

$$H_1 : x[n] = A + w[n]$$

The more general problem of detecting a given signal

$$H_0 : x[n] = w[n]$$

$$H_1 : x[n] = s[n] + w[n]$$

can be rewritten to unknown mean.

Example 1: Communication

The signal $s(t) = 2A \sin(\omega t)$ is transmitted on the carrier ω and $y(t) = s(t) + w(t)$ is received. Let

$$\begin{aligned} x &= \frac{1}{T} \int_0^T y(t) \sin(\omega t) \\ &= A + \frac{1}{T} \int_0^T w(t) \sin(\omega t) \\ &= A + w, \quad \text{under } H_1 \end{aligned}$$

where $w \sim N(0, \frac{\sigma^2}{2T})$.

Example 2: Matched filter

The signal $s[n]$ with energy E may be contained in $y[n]$. Let

$$\begin{aligned} x &= \frac{1}{N} \sum_{n=0}^{N-1} y[n]s[n] \\ &= E + w \quad \text{under } H_1 \end{aligned}$$

where $w \sim N(0, \frac{E\sigma^2}{N})$.

Neyman-Pearson

$$T(x) = \frac{\frac{1}{(2\pi\sigma^2)^{N/2}} e^{-\frac{\sum_{n=0}^{N-1}(x[n]-A)^2}{2\sigma^2}}}{\frac{1}{(2\pi\sigma^2)^{N/2}} e^{-\frac{\sum_{n=0}^{N-1}(x[n])^2}{2\sigma^2}}} = e^{-\frac{NA^2 - \sum_{n=0}^{N-1} 2Ax[n]}{2\sigma^2}} > \gamma.$$

Taking logarithm and simplifying give

$$\bar{T}(x) = \frac{1}{N} \sum_{n=0}^{N-1} x[n] > \frac{\sigma^2}{NA} \log(\gamma) + \frac{A}{2} = \bar{\gamma}.$$

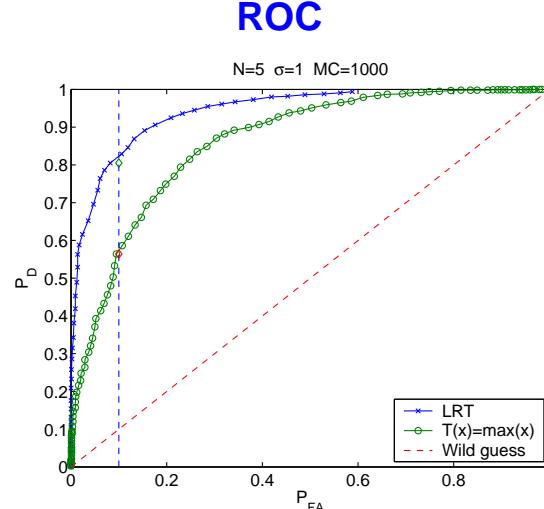
Now $\bar{T}(x) \in \begin{cases} N\left(0, \frac{\sigma^2}{N}\right) & \text{under } H_0 \\ N\left(A, \frac{\sigma^2}{N}\right) & \text{under } H_1 \end{cases}$

$$\text{so } P_{FA} = P(\bar{T}(x) > \bar{\gamma}|H_0) = Q\left(\frac{\bar{\gamma}}{\sqrt{\sigma^2/N}}\right)$$

$$\bar{\gamma} = \sqrt{\frac{\sigma^2}{N}} Q^{-1}(P_{FA})$$

$$P_D = P(\bar{T}(x) > \bar{\gamma}|H_1) = Q\left(\frac{\bar{\gamma} - A}{\sqrt{\sigma^2/N}}\right) = Q\left(Q^{-1}(P_{FA}) - \sqrt{\frac{NA^2}{\sigma^2}}\right)$$

The last expression shows that P_D is a function of the desired P_{FA} and SNR, NA^2/σ^2 , only.



Standard plots:

- **ROC:** Receiver Operating Characteristics plots P_D versus P_{FA} with $\bar{\gamma}$ being the parameter. Just guessing H_1 with probability P_{FA} gives $P_D = P_{FA}$. All clever tests give a curve above this straight line. Neyman-Pearson gives the upper bound curve for all tests.
- **Detection performance:** P_D versus SNR for a given P_{FA} . Again this performance is maximized for the Neyman-Pearson test.
- **BER:** Bit-Error Rate. In communication, both hypotheses are equal, and the design is to get $P_D = P_{FA}$. A BER-plot shows $P_D = P_{FA}$ versus SNR.

Typical Matlab code for ROC curves:

```
gamma=[-0.1:0.1:4]; MC=1000; N=5; A=1; sigma=1;
x=sigma*randn(N,MC); T1=mean(x); T2=max(x);
for k=1:length(gamma)
    PD1(k)=mean(T1+A>gamma(k));
    PD2(k)=mean(T2+A>gamma(k));
    PFA1(k)=mean(T1>gamma(k));
    PFA2(k)=mean(T2>gamma(k));
end
```

Simple line search to design γ and calculate P_D given P_{FA} :

```
PFAdesign=0.1; ind=find(PFA1<PFAdesign);
gamma1=gamma(ind(1)),
PD1design=PD1(ind(1)), ind=find(PFA2<PFAdesign);
gamma2=gamma(ind(1)), PD2design=PD2(ind(1))
interp1 better to use.
```

Maximum Likelihood Estimation

Probability density function for the observation x parameterized by the unknown parameter θ :

$$p(x, \theta) \Rightarrow \hat{\theta}_{ML} = \arg \max_{\theta} p(x_{observed}, \theta)$$

Example: If $x = (x_0 \dots x_{N-1})$, $x_i \in N(\theta, \sigma^2)$, then

$$\hat{\theta}_{ML} = \frac{1}{N} \sum_{i=0}^{N-1} x_i$$

Some Properties

Asymptotic distribution:

$$\sqrt{N}(\hat{\theta}_{ML} - \theta_0) \sim N(0, [I(\theta_0)]^{-1})$$

where the *Fisher information matrix* equals

$$I(\theta) = E \left[\frac{d}{d\theta} \log p(x, \theta) \right] \left[\frac{d}{d\theta} \log p(x, \theta) \right]^T$$

Asymptotically the best possible estimator
(reach the Cramér-Rao bound)

The GLRT for unknown A

$$\begin{aligned} H_0 : & x \in p_0(x|\theta_0) \\ H_1 : & x \in p_1(x|\theta_1) \end{aligned}$$

General principle: Plug in the maximum likelihood estimate of A under H_1 , and let

$$L(x) = \frac{\max_{\theta_1} p_1(x|\theta_1)}{\max_{\theta_0} p_0(x|\theta_0)} = \frac{p_1(x|\hat{\theta}_1^{ML})}{p_0(x|\hat{\theta}_0^{ML})}$$

Generalized Likelihood Ratio Test (GLRT)

Decide H_0 if $L(x) < \gamma$

Decide H_1 if $L(x) > \gamma$

Some more common special case:

$$\begin{aligned} H_0 : & \theta = \theta_0 \\ H_1 : & \theta \neq \theta_0 \end{aligned}$$

$$\text{GLRT becomes } L(x) = \frac{\max_{\theta_1} p(x|\theta_1)}{p(x|\theta_0)} = \frac{p(x|\hat{\theta}_1^{ML})}{p(x|\theta_0)}$$

General result for asymptotic distribution:

$$2 \log L(x) \sim \begin{cases} \chi_n^2 & \text{under } H_0 \\ \chi_n^2(\lambda) & \text{under } H_1. \end{cases}$$

$$\lambda = (\theta_1 - \theta_0)^T I(\theta_0)(\theta_1 - \theta_0)$$

where $I(\theta)$ is Fisher's information matrix

Example:

Consider

$$H_0 : x[n] = w[n]$$

$$H_1 : x[n] = A + w[n]$$

with unknown A and Gaussian noise with known variance σ^2 . Then

$$\begin{aligned} 2 \log L(x) &= \frac{N\hat{A}^2}{\sigma^2} \\ \hat{A} &= \frac{1}{N} \sum_{n=0}^{N-1} x[n] \in \begin{cases} N\left(0, \frac{\sigma^2}{N}\right) & \text{under } H_0 \\ N\left(A, \frac{\sigma^2}{N}\right) & \text{under } H_1 \end{cases} \\ \Rightarrow 2 \log L(x) &\in \begin{cases} \chi_1^2 & \text{under } H_0 \\ \chi_1^2\left(\frac{NA^2}{\sigma^2}\right) & \text{under } H_1. \end{cases} \end{aligned}$$

The MLRT for unknown A

$$H_0 : x \in p_0(x|\theta_0)$$

$$H_1 : x \in p_1(x|\theta_1)$$

General principle: Use a prior on θ and integrate out its influence in the LRT. This is the *Bayesian approach*.

$$L(x) = \frac{\int_{\theta_1} p_1(x|\theta_1)p(\theta_1)d\theta_1}{\int_{\theta_0} p_0(x|\theta_0)p(\theta_0)d\theta_0}$$

This is the **Marginalized Likelihood Ratio Test (MLRT)**.

Mixed M/G LRT approach

Test:

$$H_0 : \theta = \theta_0, \eta$$

$$H_1 : \theta \neq \theta_0, \eta$$

Here η is an unknown *nuisance parameter*. General principle using the Bayesian approach: marginalize nuisance parameters, estimate the other parameters.

The LRT becomes

$$L(x) = \frac{\max_{\theta} \int_{\eta} p(x|\theta, \eta)d\eta}{\int_{\eta} p(x|\theta_0, \eta)d\eta} = \frac{\int_{\eta} p(x|\hat{\theta}^{ML}, \eta)d\eta}{\int_{\eta} p(x|\theta_0, \eta)d\eta}$$

Nuisance parameters: classical approaches

Non-Bayesian approaches for the test

$$H_0 : \theta = \theta_0, \eta$$

$$H_1 : \theta \neq \theta_0, \eta$$

GLRT ML estimate of nuisance parameters:

$$L(x) = \frac{p(x|\hat{\theta}^{ML}, \hat{\eta}^{ML})}{p(x|\theta_0, \hat{\eta}^{ML})}$$

General result for asymptotic distribution:

$$2 \log L(x) \sim \begin{cases} \chi_{\dim(\theta)}^2 \text{ under } H_0 \\ \chi_{\dim(\theta)}^2(\lambda) \text{ under } H_1. \end{cases}$$

$$\lambda = (\theta_1 - \theta_0)^T \underbrace{(I_{\theta,\theta} - I_{\theta,\eta} I_{\eta,\eta}^{-1} I_{\eta,\theta})}_{[I^{-1}]_{\theta,\theta}} (\theta_1 - \theta_0)$$

where $I(\theta)$ is Fisher's information matrix partitioned as

$$I = \begin{pmatrix} I_{\theta,\theta} & I_{\theta,\eta} \\ I_{\eta,\theta} & I_{\eta,\eta} \end{pmatrix}$$

evaluated at the true θ_0, η_0 (also θ_1 denotes the true value under H_1).

Note that the underbraced expression $[I^{-1}]_{\theta,\theta}$ is the Schur complement to compute the upper left corner of the matrix inverse.

Rao test

Use

$$L(x) = \frac{d \log(p(x|\theta, \eta))}{d\theta} \Big|_{\theta_0, \hat{\eta}|H_0} [I^{-1}(\theta_0, \hat{\eta}|H_0)]_{\theta,\theta} \frac{d \log(p(x|\theta, \eta))}{d\theta} \Big|_{\theta_0, \hat{\eta}|H_0}$$

where $\hat{\eta}|H_0$ is the ML estimate under H_0 .

Same performance as GLRT.

Note that θ does not have to be estimated here, as it has in the Wald test.

Wald test

Use

$$L(x) = (\hat{\theta} - \theta_0)^T [I^{-1}(\hat{\theta}, \eta|H_1)]_{\theta,\theta} (\hat{\theta} - \theta_0)$$

where $\hat{\theta}, \eta|H_1$ is the ML estimate under H_1 .

Same performance as GLRT.

Locally Most Powerful (LMP) test

The Locally Most Powerful (LMP) test is a special case of the Rao test for the case of scalar parameter, no nuisance parameter and one-sided test:

$$\begin{aligned} H_0 : \theta &= \theta_0 \\ H_1 : \theta &> \theta_0 \end{aligned}$$

Take the test statistic

$$L(x) = \frac{\frac{d \log(p(x|\theta))}{d\theta} \Big|_{\theta_0}}{\sqrt{I(\theta_0)}} \sim \begin{cases} N(0, 1) \text{ under } H_0 \\ N(\sqrt{I(\theta_0)(\theta_1 - \theta_0)}, 1) \text{ under } H_1 \end{cases}$$

for small changes $\theta_1 - \theta_0 > 0$. The derivation is based on local analysis by Taylor expansion.

Bayesian extensions

Priors: Assume priors $p(H_0)$ and $p(H_1)$, respectively. Bayesian decision rule, compare the *a posteriori* distribution of each hypothesis,

$$\begin{aligned} p(H_1|x) \stackrel{H_0}{\lessgtr} p(H_0|x) &\Leftrightarrow \\ \frac{p(x|H_1)p(H_1)}{p(x)} \stackrel{H_0}{\lessgtr} \frac{p(x|H_0)p(H_0)}{p(x)} &\Rightarrow \\ \frac{p(x|H_1)}{p(x|H_0)} \stackrel{H_0}{\lessgtr} \frac{p(H_1)}{p(H_0)} = \gamma \end{aligned}$$

which shows that Bayes' rule provides a way to design the threshold in the LRT test.

Multiple tests

The use of prior enables multiple hypotheses tests:

$$\hat{k} = \arg \max_k p(H_k|x) = \arg \max_k p(x|H_k)p(H_k)$$

Bayes Risk

Define a cost to each decision

$$C_{ij} = P(H_i|H_j)$$

that is, C_{10} is the false alarm cost and $C_{0,1}$ is the missed detection cost.

Then

$$\frac{p(x|H_1)}{p(x|H_0)} \stackrel{H_0}{\lessgtr} \frac{(C_{10} - C_{00})p(H_1)}{(C_{01} - C_{11})p(H_0)} = \gamma$$

minimizes the Bayes risk or expected cost

$$\mathcal{R} = E(C) = \sum_{i=0}^1 \sum_{j=0}^1 C_{ij} P(H_i|H_j) p(H_j)$$

Linear regressions

Model: $Y = \Phi^T \theta + E$

where $\text{Cov}(E) = C$ and

$$p(Y|\theta) = \frac{1}{(2\pi)^{N/2} \sqrt{\det(C)}} e^{-\frac{1}{2}(Y - \Phi^T \theta)^T C^{-1} (Y - \Phi^T \theta)}.$$

Known parameter

Test: $\begin{cases} H_0 : \theta = 0 \\ H_1 : \theta = \theta_1 \end{cases}$

Neyman-Pearson gives LRT: $\log \left(\frac{p(Y|\theta_1)}{P(Y|\theta=0)} \right)$
 $= -0.5(Y - \Phi^T\theta_1)^T C^{-1}(Y - \Phi^T\theta_1) + 0.5Y^T C^{-1}Y$
 $= Y^T C^{-1}\Phi^T\theta_1 - 0.5\theta_1^T \Phi C^{-1}\Phi^T\theta_1$

The last term is independent of data, so take

$$L(Y) = Y^T C^{-1}\Phi^T\theta_1 \in N(\theta^T \Phi C^{-1}\Phi^T\theta_1, \theta_1^T \Phi C^{-1}\Phi^T\theta_1),$$

Constant in noise: $P_D = Q \left(Q^{-1}(P_{FA}) - \sqrt{\theta_1^T \Phi C^{-1}\Phi^T\theta_1} \right)$

Correlation detector or Matched Filter

Step 1 Pre-whiten the data:

$$\bar{Y} = C^{-1/2}Y, \quad \bar{\Phi} = C^{-1/2}\Phi \Rightarrow \\ \bar{Y} = \bar{\Phi}^T\theta + \bar{E} \quad \text{Cov}(\bar{E}) = I$$

Step 2: $L = \bar{Y}^T M$, where $M = \bar{\Phi}^T\theta_1$

Correlation detector: $L = \sum_{i=1}^N \bar{y}[i]m[i]$

Matched filter: Convolution $l(t) = \sum_{i=0}^{t-1} h[i]\bar{y}[t-i]$

Impulse response $h[i] = m[N-i]$, (matched to the signal)

Test statistics: $L = l(N)$

GLRT

Unknown parameter

Test

$$\begin{aligned} H_0 : \theta &= 0 \\ H_1 : \theta &\neq 0 \end{aligned}$$

ML estimate

$$\hat{\theta} = (\Phi\Phi^T)^{-1}\Phi Y$$

$$\begin{aligned} L(x) &= 2 \log \left(\frac{p(Y|\hat{\theta})}{P(Y|\theta=0)} \right) \\ &= -(Y - \Phi^T\hat{\theta})^T C^{-1}(Y - \Phi^T\hat{\theta}) + Y^T C^{-1}Y \\ &= 2Y^T C^{-1}\Phi^T\hat{\theta} - \hat{\theta}^T \Phi C^{-1}\Phi^T\hat{\theta} \\ &= \dots = Y^T C^{-1}\Phi^T(\Phi C^{-1}\Phi^T)^{-1}\Phi C^{-1}Y \\ &\in \chi_{\dim(\theta)}^2 \end{aligned}$$

See Chapter 6.1 in Gustafsson for more calculations on the linear model.

Exercises

- Check the lecture notes and report errors to me
- Groups of at most five persons
- I want one TEX version of problems + solution per group
- Kay: 1.1, 1.2, 1.4, 2.2, 3.1, 3.8, 3.13