

## Summary of lecture 4

- I **Signal in noise problem:**  $u(t) = s(t) + n(t)$ . Remove noise  $n(t)$  from the observed signal  $u(t)$  to obtain information about  $s(t)$ .
- II **Frequency selective filters:** Amplifies/attenuates frequencies in order to suppress noise.
- III **Stability and causality:** Poles **inside/outside** the unit circle  $\leftrightarrow$  stable and **causal/anti-causal** filter.
- IV **Linear phase filter:** Time delay is same for all frequencies.
- V **Zero phase filter:** No time delay. Non-causal, only used off-line.
- VI **Spectral factorization:** Any stationary s.p.  $y(t)$  can be realized by filtering white noise  $e(t)$  with variance  $\lambda$  through a filter  $H(q)$ ,

$$\Phi_{yy}(\omega) = \lambda \left| H(e^{i\omega}) \right|^2$$

# Outline Lecture 5

We have seen that stationary stochastic processes can be represented by

- realizations,
- covariance functions,
- spectra.

All three are of about the same complexity. In this lecture we introduce

**parametric signal models,**

which is a more compact representation.

I Standard signal sources

II AR, ARMA and state space models

III Converting transfer functions to

- ① spectrum
- ② covariance function
- ③ state space model

IV Prediction using signal models

# Signal modelling

- Any stationary s.p.  $y(t)$  can be represented as

$$y(t) = T(q)u(t)$$

where  $T(q)$  is a filter and  $u(t)$  is white noise input signal.

- Super formula in spectral domain gives

$$\Phi_{yy}(\omega) = |T(e^{i\omega})|^2 \Phi_{uu}(\omega) = |T(e^{i\omega})|^2 \lambda$$

- The filter shapes the spectrum, the input injects energy.
- Many  $u(t)$  have the property that  $\Phi_{yy}(\omega) = |T(e^{i\omega})|^2 \lambda$ . This can be used to shape the waveform  $y(t)$ .

## Standard signal sources

A signal  $y(t)$  of length  $N$  such that  $\Phi_{yy}(\omega) = |T(e^{i\omega})|^2 \lambda$ ,  $E_y = N\lambda$ .

- White noise. Most common source, often Gaussian distributed

$$u(t) \sim \mathcal{N}(0, \lambda)$$

Ex: Measurement noise, non-voiced sounds like f, v, s.

- Pulse

$$u(t) = \sqrt{N\lambda}\delta(t) = \begin{cases} \sqrt{N\lambda} & t = 0 \\ 0 & t \neq 0 \end{cases}$$

Ex: disturbances in a control systems, explosive/implosive sounds like b, d, k, t.

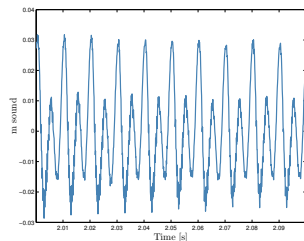
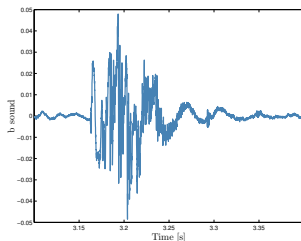
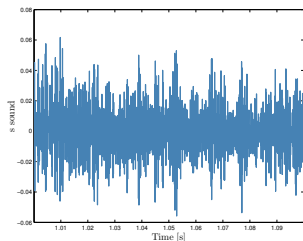
- Pulse train (note difference with book)

$$u(t) = \sqrt{P\lambda} \sum_{k=0}^{n_P} \delta(t - kP) \quad n_P = \left\lceil \frac{N}{P} \right\rceil$$

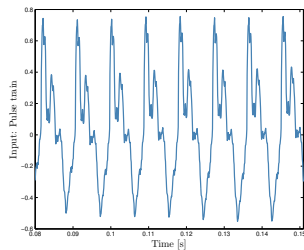
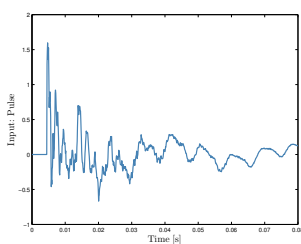
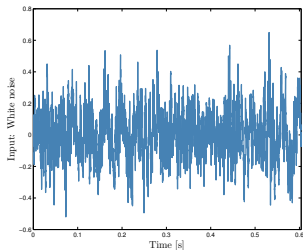
Ex: voiced sounds like m, n, l.

# Standard signal sources: examples

Recordings of s, b and m sound.



Same filter  $T(q)$  for inputs white noise, pulse and pulse train.



# State-space models: motivation

There are many good reasons for using state-space models in signal processing, e.g.:

- ① Typical result of physical model building, i.e. “easy” to construct and interpret the models.
- ② Powerful representation for analysis of the system properties.
- ③ Used in the Kalman filter
- ④ Very general model (multivariable, time varying, etc)

## Conversion example: AR to state-space

We have an AR model

$$y(t) + a_1y(t-1) + \dots + a_ny(t-n) = e(t)$$

Letting the state vector be  $x(t) = [y(t-1) \dots y(t-n)]^T$  results in

$$x(t+1) = \begin{bmatrix} -a_1 & -a_2 & \dots & -a_{n-1} & -a_n \\ 1 & 0 & \dots & 0 & 0 \\ 0 & 1 & \dots & 0 & 0 \\ \vdots & \vdots & \ddots & \vdots & \vdots \\ 0 & 0 & \dots & 1 & 0 \end{bmatrix} x(t) + \begin{bmatrix} 1 \\ 0 \\ 0 \\ \vdots \\ 0 \end{bmatrix} e(t)$$
$$y(t) = \begin{bmatrix} -a_1 & -a_2 & \dots & -a_{n-1} & -a_n \end{bmatrix} x(t) + e(t)$$

## Conversion example: ARMA to state-space (I/II)

We have an ARMA model,

$$y(t) + a_1y(t-1) + \dots + a_ny(t-n) = e(t) + c_1e(t-1) + \dots + c_me(t-m)$$

Let  $m = n$  (if not, just add zero coefficients),

$$\begin{aligned} T(q) &= \frac{1 + c_1q^{-1} + \dots + c_nq^{-n}}{1 + a_1q^{-1} + \dots + a_nq^{-n}} = 1 + \frac{(c_1 - a_1)q^{-1} + \dots + (c_n - a_n)q^{-n}}{1 + a_1q^{-1} + \dots + a_nq^{-n}} \\ &= 1 + \sum_{k=1}^n (c_k - a_k) \frac{q^{-k}}{1 + \dots + a_nq^{-n}} \end{aligned}$$

and introduce the state variables

$$x_1(t) = \frac{q^{-1}}{1 + \dots + a_nq^{-n}} e(t), \dots, x_n(t) = \frac{q^{-n}}{1 + \dots + a_nq^{-n}} e(t)$$



## Conversion example: ARMA to state-space (II/II)

Using the relationships

$$x_i(t+1) = qx_i(t) = q \frac{q^{-i}}{A(q)} e(t) = \frac{q^{-(i-1)}}{A(q)} e(t) = x_{i-1}(t)$$

and  $q^{-k}x_1(t) = x_{1+k}(t)$  we get

$$x(t+1) = \begin{bmatrix} -a_1 & -a_2 & \dots & -a_{n-1} & -a_n \\ 1 & 0 & \dots & 0 & 0 \\ 0 & 1 & \dots & 0 & 0 \\ \vdots & \vdots & \ddots & \vdots & \vdots \\ 0 & 0 & \dots & 1 & 0 \end{bmatrix} x(t) + \begin{bmatrix} 1 \\ 0 \\ 0 \\ \vdots \\ 0 \end{bmatrix} e(t)$$
$$y(t) = [c_1 - a_1 \quad c_2 - a_2 \quad \dots \quad c_n - a_n] x(t) + e(t)$$

Note that the AR model is just a special case of the above.

## Conversion: Matlab

`tf2ss` converts a transfer function  $T(q)$  to a state-space model

$$\begin{aligned}x(t+1) &= Ax(t) + Be(t) \\ y(t) &= Cx(t) + De(t)\end{aligned}$$

`ss2tf` converts a state-space model to a transfer function model

$$T(q) = C(qI - A)^{-1}B + D \quad \text{or} \quad T(z) = C(zI - A)^{-1}B + D$$

We can always make a variable change  $\tilde{x}(t) = Px(t)$ ,  $P$  invertible.  
This gives a different state space representation

$$\begin{aligned}\tilde{x}(t+1) &= PAP^{-1}\tilde{x}(t) + PBe(t) \\ y(t) &= CP^{-1}\tilde{x}(t) + De(t)\end{aligned}$$

but the same  $T(q)$ .

## Prediction for transfer function models

We have  $y(t+k) = T(q)e(t+k)$  on impulse response form,

$$y(t+k) = \sum_{i=0}^{\infty} t(i)e(t+k-i) = \sum_{i=0}^{k-1} t(i)e(t+k-i) + \sum_{i=k}^{\infty} t(i)e(t+k-i)$$

Replace the unknown future noise signal values with **zero** in predictor

$$\hat{y}(t+k|t) = \sum_{i=k}^{\infty} t(i)e(t+k-i) = \sum_{j=0}^{\infty} t(j+k)e(t-j) = \tilde{T}_k(q)e(t)$$

$y(t) = T(q)e(t) \Leftrightarrow e(t) = T^{-1}(q)y(t)$ , and thus the predictor is

$$\hat{y}(t+k|t) = \tilde{T}_k(q)T^{-1}(q)y(t)$$

## Summary of Lecture 5

- **Signal model:** A filter  $T(q)$  and a signal source  $u(t)$ .
- **Signal modelling:** The problem of finding the filter and source that makes the modeled signal as close to the true signal as possible.
- **ARMA model:** Auto-Regressive Moving Average signal model,

$$y(t) + \sum_{k=1}^n a_k y(t-k) = e(t) + \sum_{\ell=1}^m c_{\ell} e(t-\ell)$$

- **State space model:** A general and powerful way to model signals. Innovation form,

$$\begin{aligned}x(t+1) &= Ax(t) + Be(t) \\ y(t) &= Cx(t) + De(t)\end{aligned}$$

- **$k$ -step prediction:** Based on information about signal up to time  $t$ , use model to compute best prediction of signal value at time  $t+k$ ,

$$\hat{y}(t+k|t)$$