We have used the simple case of curve fitting to illustrate basic issues, Identification of Linear and Nonlinear Dynamical frameworks and techniques for linear and nonlinear system identification Systems Choice of model parametrization, model size and parameter values. Theme 2: Linear Models Parametric – Nonparametric methods Parameter values easy: Some version of least squares fit. Basic asymptotic properties: $\hat{\theta}_N \rightarrow \theta^*$, best possible approximation Lennart Ljung available in the parameterization (for the used x_t -sequence) $\sqrt{N}(\hat{\theta}_N - \theta^*) \sim N(0, P), P = \lambda [E\psi(t)\psi^T(t)]^{-1}$ (Normal distribution) **Division of Automatic Control** Choice of parametric model structure guided by bias-variance trade off Linköping University (number of parameters) Sweden Choice of nonparametric method guided by bias-variance trade off (bandwidth of the kernel) System Identification: Linear Models Lennart Ljung System Identification: Linear Models Lennart Ljung Berkelev, 2005 Berkeley, 2005 Today: Goal Focus in This Theme Goal: Estimate a linear model in discrete or continuous time with or without 1. Frequency response function (FRF) an additive noise model 2. Data in time and frequency domain 3. The richness of parameterizations of linear models $y(t) = G(\sigma)u(t) + v(t)$ 4. Fitting parameterized linear models to data Time domain data σ is differentiation operator p or shift operator q. Frequency domain data The corresponding frequency response function (FRF) is $G(i\omega)$ or $G(e^{i\omega T})$. 5. Noise models Estimating a linear system is the same as estimating its FRF-curve. 6. The asymptotic properties of the estimates 7. Subspace methods 8. Spectral Analysis AUTOMATIC CONTROL COMMUNICATION SYSTEMS LINKÖPINGS UNIVERSITET System Identification: Linear Models Lennart Ljung Berkeley, 2005 System Identification: Linear Models Lennart Ljung Berkeley, 2005 The Frequency Response Function, FRF **Background Material** Linear differential (difference) equations A linear system is characterized by its transfer function G(s) (the Laplace transform of its impulse response) ■ Linear system, State-space model, Transfer function, Evaluated on the imaginary axis, this gives the FRF $G(i\omega)$, which describes Frequency response function, Bode plot the response to sinusoidal inputs: Fourier transform, DFT, Signal spectrum $u(t) = A\cos(\omega t), \quad y(t) = A_1\cos(\omega t + \phi)$ Noise representation, Kalman Filter $A_1 = |G(i\omega)|A, \quad \phi = \arg G(i\omega)$ This could be a way of determining G (frequency analysis). All Frequencies at the same time: $Y(i\omega) = G(i\omega)U(i\omega)$ Y and U are the Fourier transforms of the output and input. System Identification: Linear Models Lennart Ljung System Identification: Linear Models Berkelev, 2005 Berkelev, 2005 Lennart Liung The Bode Plot Focus in This Theme 10 1. Frequency response function (FRF) 4 Amplitude 10⁰ 2. Data in time and frequency domain 3. The richness of parameterizations of linear models 4. Fitting parameterized linear models to data Time domain data 10 10 10 10 10 Frequency domain data 5. Noise models 6. The asymptotic properties of the estimates -50 (degrees) 7. Subspace methods -100 8. Spectral Analysis Phase (-150 -200

Summary Theme 1: Curve Fitting

-250 10

System Identification: Linear Models

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Frequency (rad/s)

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Data from Dynamic Systems: Input-Output Data

- Discrete time
 - Time-domain: $\{u(1), y(1), u(2), y(2), \dots, u(N), y(N)\}$
 - Frequency-domain $\{U_N(e^{i\omega_1}), Y_N(e^{i\omega_1}), \dots, U_N(e^{i\omega_N}), Y_N(e^{i\omega_N})\}$ DFT-grid: $\omega_k = 2\pi k/N$

$$U_N(z) = \frac{1}{\sqrt{N}} \sum_{k=1}^N u(k) z^{-k}$$

- Continuous time
 - Frequency-domain $\{U_N(i\omega_1), Y_N(i\omega_1), \dots, U_N(i\omega_N), Y_N(i\omega_N)\}$

$$U_N(s) = \frac{1}{\sqrt{N}} \int_0^N u(t) e^{-st} dt$$

(Band limited, periodic data)

- Frequency Response Data (FRD)
 - From frequency analyzers or computed/estimated using FFT techniques
 - $\hat{G}(i\omega_k)$ or $\hat{G}(e^{i\omega_k}), k = 1, 2, \dots, N$
 - Possibly with uncertainty measures $W(i\omega_k)$
 - Simple estimate, ETFE:

$$\hat{\hat{G}}_{N}(i\omega) = \frac{Y_{N}(i\omega)}{U_{N}(i\omega)} \qquad \text{Variance: } W(i\omega) = \frac{\Phi_{v}(\omega)}{|U_{N}(i\omega)|^{2}}$$

where $\Phi_v(\omega)$ is the spectrum of the output disturbance

• Other estimates (spectral analysis): smoothed versions of ETFE (more later)

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ETFE: Empirical Transfer Functio	on Estim	ate 11	Focus in This Theme		
10^2 10^1 10^0 10^{-1} 10^{-1} 10^0 Frequency (rad/s)	io'		 Frequency response function (FRF) Data in time and frequency domain The richness of parameterizations of line Fitting parameterized linear models to da Time domain data Frequency domain data Noise models The asymptotic properties of the estimate Subspace methods Spectral Analysis 	ar models ta es	
System Identification: Linear Models Lennart Ljung	Berkeley, 2005	AUTOMATIC CONTROL COMMUNICATION SYSTEMS LINKÓPINGS UNIVERSITET	System Identification: Linear Models Lennart Ljung	Berkeley, 2005	AUTOMATIC CONTROL COMMUNICATION SYSTEMS LINKÖPINGS UNIVERSITET
Linear Model Structures			Linear Model Structures, cont'o		
Model Structure = Model Parameterization First without noise model $y(t) = G(q, \theta)u(t), \qquad q \text{ shift operator}$ FIR: $y(t) = b_1u(t-1) + b_2u(t-2) - y(t) = B(q)u(t), \theta = [b_1, \dots, B(q) = b_1q^{-1} + \dots + b_nq^{-n}$ $B(q) = b_1q^{-1} + \dots + b_nq^{-n}$ OE: $y(t) = \frac{b_1q^{-1} + \dots + b_nq^{-n}}{1 + f_1q^{-1} + \dots + f_nq^{-n}}u(t)$ Linear difference equation $y(t) + f_1y(t-1) + \dots + f_ny(t-n) = b_1$	$r: qu(t) = u(t)$ $+ \dots + b_n u(t)$ $b_n]^T$ $u(t-1) + \dots$	(t+1) (t-n) (t-n)	State space: $x(t+1) = A(\theta)x(t) + B(t)$ $y(t) = C(\theta)x(t) + D(\theta)wt$ $G(q, \theta) = C(\theta)[qI - F(\theta)]$ The matrices can be arb Process model: Static gain, times $T_p\dot{y}(t) + y(t) = t$ $G(s, \theta) = \frac{F}{1+t}$	$ \begin{array}{l} \theta) u(t) \\ (t) \\]^{-1} B(\theta) + D(\theta) \\ \text{itrarily paramete} \\ \text{ne constant, del} \\ = K u(t-T_d) \\ \frac{\zeta}{sT_p} e^{-T_d s} \end{array} $	rized by θ ay
System Identification: Linear Models Lennart Ljung	Berkeley, 2005	AUTOMATIC CONTROL COMMUNICATION SYSTEMS LINKÖPINGS UNIVERSITET	System Identification: Linear Models Lennart Ljung	Berkeley, 2005	AUTOMATIC CONTROL COMMUNICATION SYSTEMS LINKÖPINGS UNIVERSITET
Parameterizations of Linear Dyna	amic Mo	dels 15	Focus in This Theme		
A Linear Model Structure is (one way or another) parameterization of the frequency function $G(e^{i\omega}, \theta) \text{or} G(i\omega, \theta)$ Typical Cases: $(x = e^{i\omega} \text{ or } x = i\omega)$ $\text{FIR: } G(x, \theta) = \theta_1 + \theta_2 x + \ldots + \theta_n x^{n-1}$ $\text{OE: } G(x, \theta) = \frac{\theta_1 + \theta_2 x + \ldots + \theta_n x^{n-1}}{1 + \theta_{n+1} x + \ldots + \theta_n x^{n-1}}$ $\text{Laguerre and similar: } G(x, \theta) = \sum_{k=1}^{N} \theta_k L_k(x)$ $\text{Via state space: } G(x, \theta) = C(\theta)(xI - A(\theta))^{-1}B(\theta) + $ $\text{Parameterization of the state space matrices } A, B$ $\text{Local parameterization: } G(x, \theta) \text{ piecewise constant over versions thereof}.$ $\text{Process Models: } G(i\omega, \theta) = \frac{K}{1 + i\omega T_p} e^{-i\omega T_d} \theta = [K, R]$	$D(\theta)$ C, D: Physical er a frequency g $T_p, T_d]$	I, free or canonical rrid (or smoothed	 Frequency response function (FRF) Data in time and frequency domain The richness of parameterizations of line Fitting parameterized linear models to da Time domain data Frequency domain data Noise models The asymptotic properties of the estimate Subspace methods Spectral Analysis 	ar models ta es	
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Bottom line curve-fitting: $y(k) = g(x_k) + v_k, E|v_k|^2 = \lambda_k$

Time domain: $y(t) = G(q, \theta)u(t) + v(t) Ev^2(t) = \lambda$

 $Y_k = G(i\omega_k, \theta)U_k + V_k E|V_k|^2 = \lambda, \quad (Y_k = Y(i\omega_k))$

 $\min \sum |y(k) - g(x_k, \theta)|^2 / \lambda_k$

 $\min \sum |y(t) - G(q,\theta)u(t)|^2$

 $\min\sum |Y_k - G(i\omega_k, \theta)U_k|^2$

 $\min \sum |\hat{\hat{G}}_k - G(i\omega_k, \theta)|^2 / W_k$

FRD: $\hat{G}_k = G(i\omega_k, \theta) + \tilde{V}_k$

Frequency domain:

- Input-Output Data, Discrete time, DFT frequency grid:

$$\sum |y(t) - G(q,\theta)u(t)|^2 \iff \sum |Y_k - G(e^{i\omega_k},\theta)U_k|^2$$

Parseval ■ FRD being ETFE for the DFT frequency grid:

$$\begin{split} \hat{\hat{G}}_k &= \frac{Y_k}{U_k}, \quad W_k = \frac{\Phi_v(\omega_k)}{|U_k|^2} \quad (\Phi_v(\omega) = \lambda) \\ &\sum |\hat{\hat{G}}_k - G(e^{i\omega_k}, \theta)|^2 / W_k = \sum |Y_k - G(e^{i\omega_k}, \theta)U_k|^2 / \lambda \end{split}$$

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Pre-filtering: A Further Degree of	Freedon	n ₁₉	Focus in Th	is Theme	

Pre-filter the data before the fit: $y_F(t) = L(q)y(t)$, $u_F(t) = L(q)u(t)$. Gives

$$\sum |y_F(t) - G(q, \theta)u_F(t)|^2 = \sum |L(q)(y(t) - G(q, \theta)u(t))|^2$$

$$\sim \sum |L(e^{i\omega_k})|^2 |Y_k - G(e^{i\omega_k}, \theta)U_k|^2$$

 $E|\tilde{V}_k|^2 = W_k$

"Relevance weighting" in curve fitting. The fit is focused to the frequency ranges where L is large. Think of L as a band-pass filter.

- 1. Frequency response function (FRF)
- 2. Data in time and frequency domain
- 3. The richness of parameterizations of linear models
- 4. Fitting parameterized linear models to data Time domain data
 - Frequency domain data
- 5. Noise models
- 6. The asymptotic properties of the estimates
- 7. Subspace methods
- 8. Spectral Analysis

System Identification: Linear Models Berkeley, 2005 Lennart Ljung	AUTOMATIC CONTROL COMMUNICATION SYSTEMS LINKÖPINGS UNIVERSITET	System Identification Lennart Ljung	n: Linear I	Models		Berkeley, 2005	AUTOMATIC CONTROL COMMUNICATION SYSTEMS LINKÖPINGS UNIVERSITET
Linear Models with Noise		Some T	ypic	al	Model Structures		
Assume that the output is corrupted by additive noise with kn	own or unknown	Acronym	G	H	Equation		
properties.		ARX	$\frac{B}{A}$	$\frac{1}{A}$	A(q)y(t) = B(q)u(t) + e(t)		
$y(t) = G(q, \theta) + v(t)$		OE	$\frac{B}{F}$	1	$y(t) = \frac{B(q)}{F(q)}u(t) + e(t)$		
$v(t)$ has spectrum $\Phi_v(\omega, \theta) = \lambda H(e^{\imath \omega}, \theta) ^2$		ARMAX	$\frac{B}{A}$	$\frac{C}{A}$	A(q)y(t) = B(q)u(t) + C(q)e	(t)	
Same as $c_{1}(t) = C(a, \theta)c_{2}(t) + H(a, \theta)c_{3}(t)$		BJ	$\frac{B}{F}$	$\frac{C}{D}$	$y(t) = \frac{B(q)}{F(q)}u(t) + \frac{C(q)}{D(q)}e(t)$		
g(t) = G(q, b)a(t) + H(q, b)e(t) where <i>e</i> is white noise with variance λ $H(q, \theta)$ is monic, i.e. $H(0, \theta) = 1$.		State-Space: $G(a, \theta) = C(\theta)(aI - F(\theta))^{-1}B(\theta), H(a, \theta) = C(\theta)(aI - F(\theta))^{-1}B(\theta),$			$-C(\theta)(aI -$	$F(\theta))^{-1}K(\theta) + I$	
Frequency Domain:		x(t +	1) = 1	$F(\theta)$	$x(t) + B(\theta)u(t) + K(\theta)e(t)$	= C (V)(q1	I(0)) I(0) + I
$V = O(-i\psi_{E} - 0)U + V = E[V]^{2} = \Phi(-, 0) (V$	$\mathbf{v}(i\omega_{k})$	y((t) = 0	$C(\theta)$	x(t) + e(t)		
$I_k = G(e^{-\kappa}, \theta)U_k + V_k, L V_k = \Phi(\omega_k, \theta), (I_k =$	r (e))	Comm	non or	diffe	erent dynamics in input and no	ise channels	6.
System Identification: Linear Models Berkeley, 2005 Lennart Ljung	AUTOMATIC CONTROL COMMUNICATION SYSTEMS LINKÖPINGS UNIVERSITET	System Identification Lennart Ljung	n: Linear I	Models		Berkeley, 2005	AUTOMATIC CONTROL COMMUNICATION SYSTEMS LINKÖPINGS UNIVERSITET
One step ahead prediction for models with	th noise 23	Check t	he p	ore	dictor formula		
				- 1/			
Assume there is a delay in $G(q)$		$\hat{y}(t t-1) =$ 1 White a	[1 - I]	$H^{-1}($	$[q]y(t) + H^{-1}(q)G(q)u(t)$	P(a)u(t)	
y(t) = G(q)u(t) + H(q)e(t)		2. ARX m	odel:	(e wl	hite) $y(t_1, y_2(t_1, t_2)) = 0$	r(q)u(v)	
$H (q)y(t) = H (q)G(q)u(t) + e(t)$ $u(t) = [1 - H^{-1}(q)]u(t) + H^{-1}(q)G(q)u(t) + H^{-1}(q)G(q)u(t)$	+ e(t)	$y(t) + a_1y(t-1) + \ldots + a_ny(t-1) = b_1u(t)$				$(t-1) + \ldots +$	$+ b_n u(t-n) + e(t)$
Since H is monic, $1 - H^{-1}(0) = 0$ so	· ·	$A(q)y(t) = B(q)u(t) + e(t), G(q) = \frac{B(q)}{A(q)}, H(q) = \frac{1}{A(q)}$					
$[1 - H^{-1}(q)]y(t) = \tilde{h}_1 y(t-1) + \tilde{h}_2 y(t-2) + .$			$[1 - H^{-}]$	$-H^{-}$	$[1(q)] = -a_1 q^{-1} - \dots - a_n q^{-n}$ G(q) = B(q)	(1)	

so the RHS of the above expression is actually known at time t-1. Since e(t) is unpredictable at time t-1 the predictor must be

 $\hat{y}(t|t-1) = [1-H^{-1}(q)]y(t) + H^{-1}(q)G(q)u(t)$

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$$\begin{aligned} y(t) + a_1 y(t-1) + \ldots + a_n y(t-1) &= b_1 u(t-1) + \ldots + b_n u(t-n) + e(t) \\ A(q) y(t) &= B(q) u(t) + e(t), \quad G(q) = \frac{B(q)}{A(q)}, \\ H(q) &= \frac{1}{A(q)} \\ & [1 - H^{-1}(q)] = -a_1 q^{-1} - \ldots - a_n q^{-n} \\ H^{-1}(q) G(q) &= B(q) \\ \hat{y}(t|t-1) &= -a_1 y(t-1) - \ldots - a_n y(t-n) + b_1 u(t-1) + \ldots + b_n u(t-n) \end{aligned}$$

3. Kalman filter ...

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Frequency domain: $Y_{k} = G(e^{i\omega_{k}})U_{k} + V_{k} E V_{k} ^{2} = \Phi(\omega_{k}) = \lambda H(e^{i\omega_{k}}, \theta) ^{2}$ $\min \sum Y_{k} - G(e^{i\omega_{k}}, \theta)U_{k} ^{2}/\Phi(\omega_{k})$ $= \min \sum H^{-1}(e^{i\omega_{k}}, \theta)[Y_{k} - G(e^{i\omega_{k}}, \theta)U_{k}] ^{2}$ (7. Subspace m 8. Spectral Ana	otic properties of the estimates nethods alysis	
System Identification: Linear Models AutoAntr.common, Lennart Ljung System Identification: Linear Mod	dels Berkeley.2005 AUTOMATIC CONTROL COMMUNICATION STRAIN LINKÖPINGS LINKERSTET	

- Assume that data have been generated by $y(t) = G_0(q)u(t) + H_0(q)e(t)$, where e is white noise with variance λ .

Tra

$$\hat{\theta}_N = \arg\min \sum_{i} |L(q)\varepsilon(t,\theta)|^2$$
$$\varepsilon(t,\theta) = H^{-1}(q,\theta)[y(t) - G(q,\theta)u(t)]$$

- Properties of the estimate $\hat{\theta}_N$
- A parameter-free assessment of model quality for linear systems and models
- How to affect the model quality?

			Also, asymptotic normality of $\sqrt{N}(b)$	$\theta_N - \theta^*)$	
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Quality in The Frequency Domai	n: Bias		A Paradox		

True system: $y(t) = G_0(q)u(t) + v(t)$ $\text{Model:} \ \hat{G}_N(e^{i\omega}) = G(e^{i\omega}, \hat{\theta}_N), \quad \hat{H}_N(e^{i\omega}) = H(e^{i\omega}, \hat{\theta}_N)$

Instate properties of
$$\hat{\theta}_N$$
 to $\hat{G}_N \& \hat{H}_N$
 $\hat{G}_N(e^{i\omega}) \to G^*(e^{i\omega}), \quad \hat{\theta}_N \to \theta^* \quad \text{as } N \to \infty$

$$\theta^* \approx \arg\min \int_{-\pi}^{\pi} \mid G_0(e^{i\omega}) - G(e^{i\omega}, \theta) \mid^2 \cdot \frac{\Phi_u(\omega) \cdot \mid L(e^{i\omega}) \mid^2}{\mid H(e^{i\omega}, \theta^*) \mid^2} d\omega$$

$$\begin{split} G^*(e^{i\omega}) \text{ is closest to } G_0(e^{i\omega}) \text{ in the norm } Q(\omega) &= \frac{\Phi_u(\omega)|L(e^{i\omega})|^2}{|H(e^{i\omega},\theta^*)|^2} \\ \Phi_u: \text{ Input spectrum,} \qquad L: \text{ Pre-filter} \end{split}$$

A high order system is approximated in the ARX-structure

1. $\hat{\theta}_N \to \theta^* = \arg \min \operatorname{E}[L(q)\varepsilon(t,\theta)]^2$

If $\varepsilon(t, \hat{\theta}_N) \approx$ white noise then

Cov $\hat{\theta}_N \approx \frac{\lambda}{N} [\mathbf{E} \psi(t) \psi^T(t)]^{-1}$

 $\psi(t) = \frac{d}{d\theta} \hat{y}(t|\theta)$ d-dimensional vector

NB! This covariance can be estimated!

true system" "Best possible":

 $\lambda = \mathbf{E}\varepsilon^2(t, \hat{\theta}_N)$

$$(1 + a_1q^{-1} + a_2q^{-2})y(t) = (b_1q^{-1} + b_2q^{-2})u(t) + e(t)$$

"The estimate converges to the best possible approximation of the

0+)

and in the OE-structure

$$y(t) = \frac{b_1 q^{-1} + b_2 q^{-2}}{1 + f_1 q^{-1} + f_2 q^{-2}} u(t) + e(t)$$

Can you explain the difference?



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L is chosen as a low-pass filter

m=arx(data,[2 2 1],'focus',[0 0.2])





 $G(e^{i\omega}, \hat{\theta}_N) = \hat{G}_N(e^{i\omega})$ by

Gauss' Approximation Formula: Cov $f(\hat{\theta}_N) \approx f'$ Cov $\hat{\theta}_N (f')^T$



$$Y(t) = \begin{bmatrix} x(t+) \\ y(t) \end{bmatrix}$$
$$\Phi(t) = \begin{bmatrix} x(t) \\ u(t) \end{bmatrix}$$

System Identification: Linear Models Lennart Ljung estimated using the Least Squares method. With the covariance matrix of ν ,

the optimal Kalman gain could then be computed.

Fact: All (interesting) states can be found as linear combinations of the *k*-step ahead predictors $\hat{y}(t + k|t)$, k = 1, ..., n (the predicted value of y(t + k) based on input-output data up to time *t*. No prediction of the effect of inputs after time *t*.)

So estimate these k-step ahead predictors using ARX-models, and determine from these the good linear combinations to form the states x.

Use these x to form the linear regression to estimate A, B, C, D.

- State space basis selected automatically
- \blacksquare Form sample covariances of y and u: One SVD and one QR-step
- No iterations
- Quality properties not fully understood

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More Formal Calculations, 1/5		43 More Formal Calcul	ations, 2/5 44
$\begin{split} y(t+k) &= \sum_{j=-\infty}^{t+k} h^u_{t+k-j} u(j) + h\\ \hat{y}(t+k t) &= \sum_{j=-\infty}^t h^u_{t+k-j} u(j) + h\\ \end{split}$ Let $Y^x(t) &= \begin{bmatrix} \hat{y}(t+1 t)\\ \vdots\\ \hat{y}(t+n t) \end{bmatrix}. \end{split}$	$e_{t+k-j}^{e}(j)$ $e_{t+k-j}^{e}(j)$	So, all (Kalman) states $x(t)$, as linear combinations of Y for some L . The (minimal) of $f^{x}(t), t = 1, \ldots, N$ So, with $f^{x}(t), t = 1, \ldots, N$ conditioned. This includes the to $Y^{N} =$ Once $x(t), t = 1, \ldots, N$ have state-space matrices.	in any state-space representation can be written $x(t) = LY^x(t)$ order of the state-space representation is the rank given, pick <i>L</i> , so that $x(t)$ becomes well he choice of dimension of <i>x</i> . Typically, apply SVD $= [Y^x(1) Y^x(2) Y^x(N)]$ the been determined, proceed as above to find the
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More Formal Calculations, 3/5		45 More Formal Calcul	ations, 4/5 46
How to estimate the predictors: $y(t+k) = \sum_{i=1}^{t+k} \ h^u_{t+k-j} u(j) + h^i_t$	$_{+k-j}e(j)(*)$	$y(t+k) = \sum_{j=-\infty}^{t+k} \tilde{h}_{t+k}^u$	$-ju(j) + \sum_{j=-\infty}^{t} \tilde{h}_{t+k-j}^{e}y(j) + \sum_{j=t+1}^{t+k} h_{t+k-j}^{e}e(j)$

$$y(t+k) = \sum_{j=-\infty}^{++\kappa} h^u_{t+k-j} u(j) + h^e_{t+k-j} e(j)(*)$$
$$\hat{y}(t+k|t) = \sum_{j=-\infty}^{t} h^u_{t+k-j} u(j) + h^e_{t+k-j} e(j)$$

e(t) and y(t) have an invertible relationship.

į

$$y(t+k) = \sum_{j=-\infty}^{t+k-1} \tilde{h}^u_{t+k-j} u(j) + \tilde{h}^e_{t+k-j} y(j) + e(t+k) + \tilde{h}^u_0 u(t+k) (**)$$

so replace e(j) in (*) by y and u from (**): ...

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The Subspace Method Algorithm		47	Nonparametric Curve Fitting		48

The essence of the subspace methods is as follows

- 1. Select n and n_1 and estimate $Y^x(t), t = 1, \ldots, N$.
- 2. Determine a good choice of L in $\vec{x(t)} = LY^x(t)$ (including dimension) using SVD or similar decomposition
- 3. Possibly determine \boldsymbol{n} by visual inspection of the singular values in the above expression.
- 4. Estimate A, B, C and D by least squares in the state-space model, treating x(t) as a measured sequence.
- 5. Use the covariance matrix of ν to compute the Kalman Filter gain K.

Smoothing bottom line:

$$\hat{g}_N(x) = \sum_{k=1}^N C(x, x_k) y(k), \quad \sum_{k=1}^N C(x, x_k) = 1 \; \forall x$$

Now, truncate the first equation at $j = t - n_1$ rather than at $j = -\infty$, and

estimate \tilde{h} using the least squares method. Use these estimates in the second equation to estimate \hat{y} . The value n_1 corresponds to the "auxiliary order". All of this can be done numerically efficient by projections.

 $\mbox{Often} \quad C(x,x_k) = \mathcal{N}\tilde{c}(x-x_k)/\lambda_k \mbox{ and } \tilde{c}(r) = 0 \mbox{ for } |r| > \gamma, \quad \gamma = \mbox{the "bandwidth"}$

Curve fitting of the ETFE: $y(k) = \hat{\hat{G}}_k = \frac{Y_k}{U_k}, \quad \lambda_k = \frac{\Phi_v}{|U_k|^2}, \quad Y_k = Y(i\omega_k)$

$$\hat{G}(i\omega) = \frac{\sum \tilde{c}(\omega - \omega_k)Y_k U_k^*}{\sum \tilde{c}(\omega - \omega_k)|U_k|^2}$$

Bandwidth = Frequency Resolution - Could be frequency dependent

$$(C(\omega, \omega_k) = \tilde{c}(\omega - \omega_k, \omega))$$

Spectral Analysis (with Frequency Dependent Resolution),

 $\hat{y}(t+k|t) = \sum_{j=-\infty}^t \tilde{h}^u_{t+k-j} u(j) + \tilde{h}^e_{t+k-j} y(j)$

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- Rich possibilities to parameterize linear models (their FRFs)
- Curve fitting to parameterized FRFs
- Limit model is a weighted fit between the true FRF and the model one
- Variance of estimated FRF is approximatively proportional to noise-to-signal ratio frequency by frequency
- Subspace methods can be seen as a two step process: Estimate the states and find the state space matrices by linear regression
- Nonparametric curve fitting is the spectral analysis method

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