

Filtering and Identification

Day 3 - Lecture 2:

The optimal predictor for input-output (i/o) models and Prediction Error Optimization

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Overview

- **Recap “goals” of the course**
- The Output-error Identification “recipe”
- The prediction-error Identification “recipe”
- Case Study: the acoustical duct

Recap “goal” of the course

- **Estimate** linear static models from data:

$$\min_x \|y - Fx\|_W^2$$

Relevant and to introduce the numerical fundamentals of this course!

- **Estimate** the state of an LTI state space model:

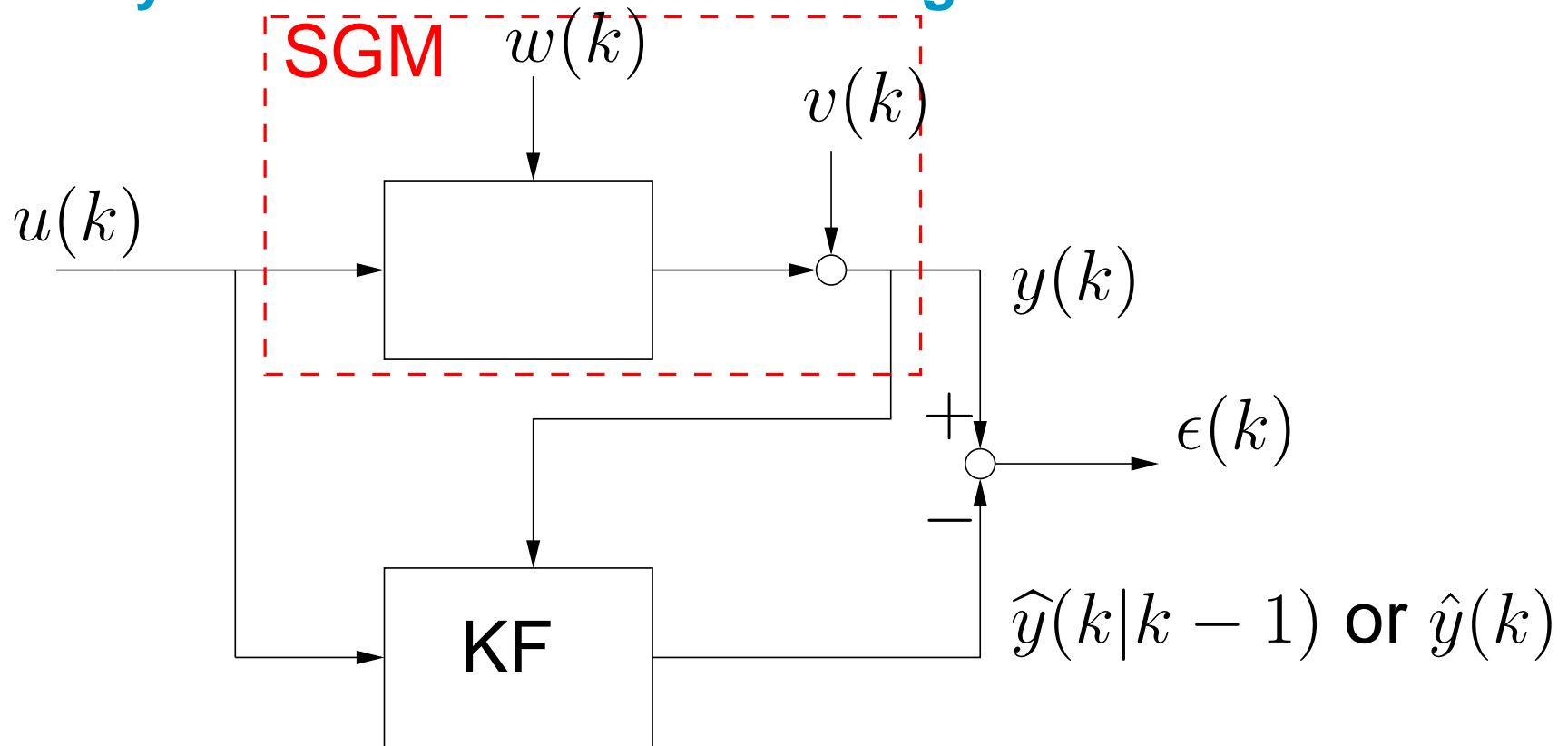
$$x(k+1) = Ax(k) + Bu(k) + Ke(k) \quad e(k) \sim (0, R_e)$$

$$y(k) = Cx(k) + Du(k) + e(k)$$

given the model (A, B, C, D) and K and input-output data.

- **“Identify”** state space model matrices (A, B, C, D, K) from input-output data $\{u(k), y(k)\}_{k=1}^N$.

The system identification challenge



1. Generally **SGM** is not known!
2. Then we have an “identification problem”
3. First we study the challenges with “classical” PEM

The “classical” identification “recipe”

1. Assume a **SGM** — say $\widehat{\text{SGM}}(\theta)$
2. Derive from $\widehat{\text{SGM}}(\theta)$ the “optimal” prediction of the output — denoted as $\hat{y}(k|k-1; \theta)$ (in short $\hat{y}(k, \theta)$)
3. Find the “best” estimate $\hat{\theta}$ by minimizing the cost function:

$$\frac{1}{N} \sum_{k=1}^N (y(k) - \hat{y}(k, \theta))^2$$

w.r.t. θ .

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The OE Identification recipe

1. Given the “true” **SGM**:

$$y(k) = G(q)u(k) + v(k) \in \mathbb{R} \quad E[u(k)v(\ell)^T] = 0$$

2. **Model** $\widehat{\text{SGM}}(\theta)$: $y(k, \theta) = \underbrace{\frac{B(q, b_i)}{A(q, a_i)}}_{z(k)} u(k) + e(k) \quad e(k) \sim (0, R)$

with $B(q, b_i) = b_1 q^{\boxed{-n_k}} + \dots + b_{n_b} q^{\boxed{-n_b - n_k + 1}}$ and

$A(q, a_i) = \boxed{1} + a_1 q^{-1} + \dots + a_{n_a} q^{\boxed{-n_a}}$ ($\boxed{\text{monic}}$) and q is

the time shift operator.

Generally $n_k = 1$ and $n_a = n_b$.

with $\theta = \left[a_i, b_i, \dots \right]$.

Finding “optimal” predictor using SS innovation form

Example OE model 3rd-order (c'td):

$$z(k) = \underbrace{-a_1 z(k-1) + b_1 u(k-1)}_{x_2(k-1)} \underbrace{-a_2 z(k-2) + b_2 u(k-2)}_{x_3(k-2)} \underbrace{-a_3 z(k-3) + b_3 u(k-3)}_{x_1(k)}$$

The definition of the states yields the following SS model of the OE $\widehat{\text{SGM}}(\theta)$

$$\begin{bmatrix} x_1(k+1) \\ x_2(k+1) \\ x_3(k+1) \end{bmatrix} = \underbrace{\begin{bmatrix} -a_1 & 1 & 0 \\ -a_2 & 0 & 1 \\ -a_3 & 0 & 0 \end{bmatrix}}_{A_3(a)} \begin{bmatrix} x_1(k) \\ x_2(k) \\ x_3(k) \end{bmatrix} + \underbrace{\begin{bmatrix} b_1 \\ b_2 \\ b_3 \end{bmatrix}}_{B_3(b)} u(k)$$

$$y(k) = \underbrace{\begin{bmatrix} 1 & 0 & 0 \end{bmatrix}}_{C_3} x(k) + e(k)$$

What is the optimal predictor of the output?

The “optimal” predictor of an output-error model

The **innovation** form of the OE-SGM model reads for

$$\theta^T = \begin{bmatrix} a^T & b^T & x(0)^T & \text{vec}(R)^T \end{bmatrix}:$$

$$x(k+1, \theta) = A(a)x(k, \theta) + B(b)u(k)$$

$$y(k, \theta) = Cx(k, \theta) + e(k) \quad e(k) \sim (0, R)$$

The Kalman filter analysis directly shows that the stationary optimal predictor equals:

$$\hat{x}(k+1, \theta) = A(a)\hat{x}(k, \theta) + B(b)u(k)$$

$$\hat{y}(k, \theta) = C\hat{x}(k, \theta)$$

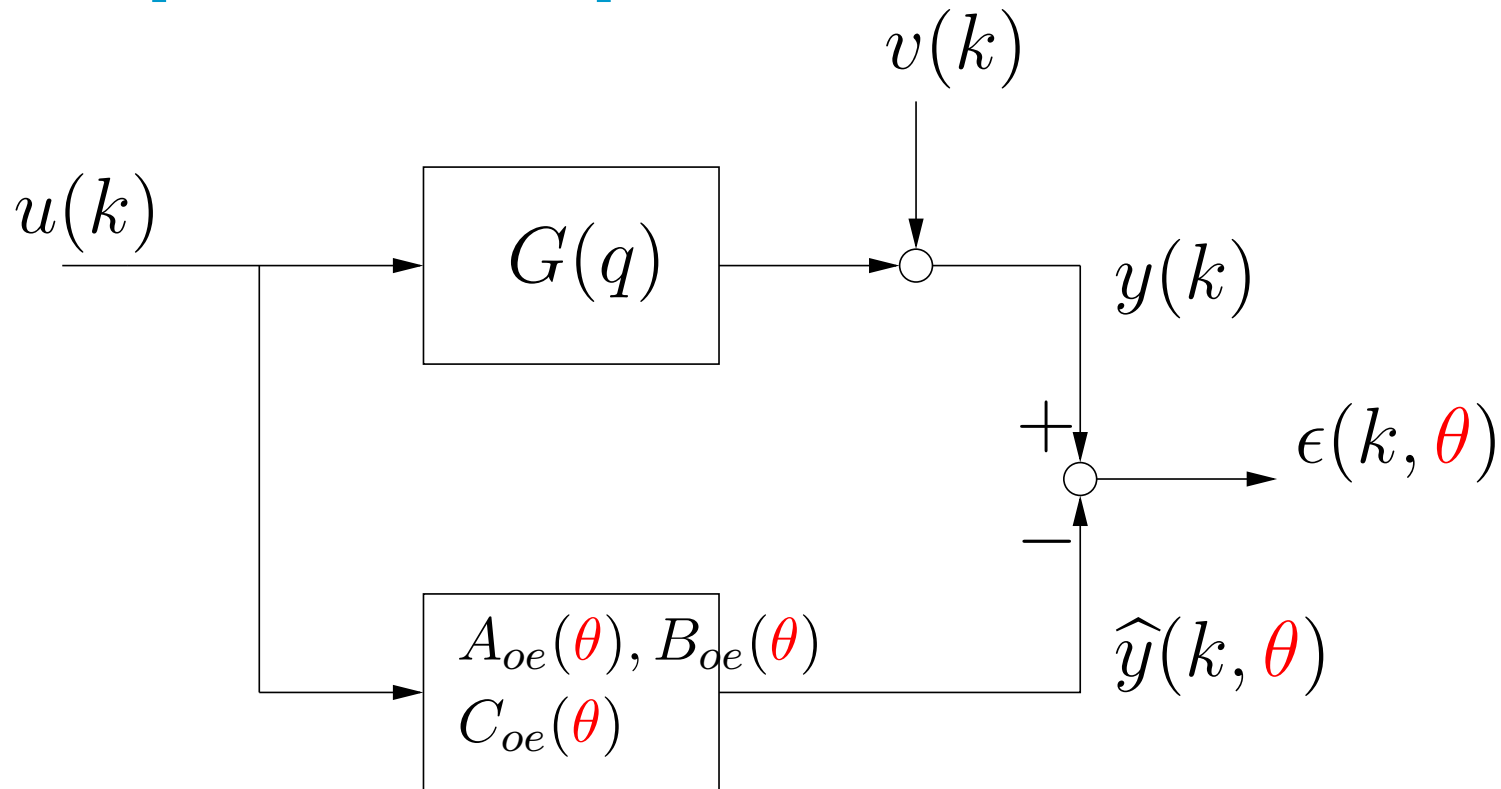
The stationary Kalman filter

The stationary solution for an LTI system is characterized by the DARE, which for the given OE model reads:

$$P = APA^T - APC^T(CPC^T + R)^{-1}CPA^T$$

The solution that satisfies this equation is $P \equiv 0$ and hence the KF gain K is also zero!

Output-error problem



$$\hat{\theta}_N = \operatorname{argmin} J_N(\theta) \quad \text{with} \quad J_N(\theta) = \frac{1}{N} \sum_{k=1}^N \|y(k) - \hat{y}(k, \theta)\|_2^2$$

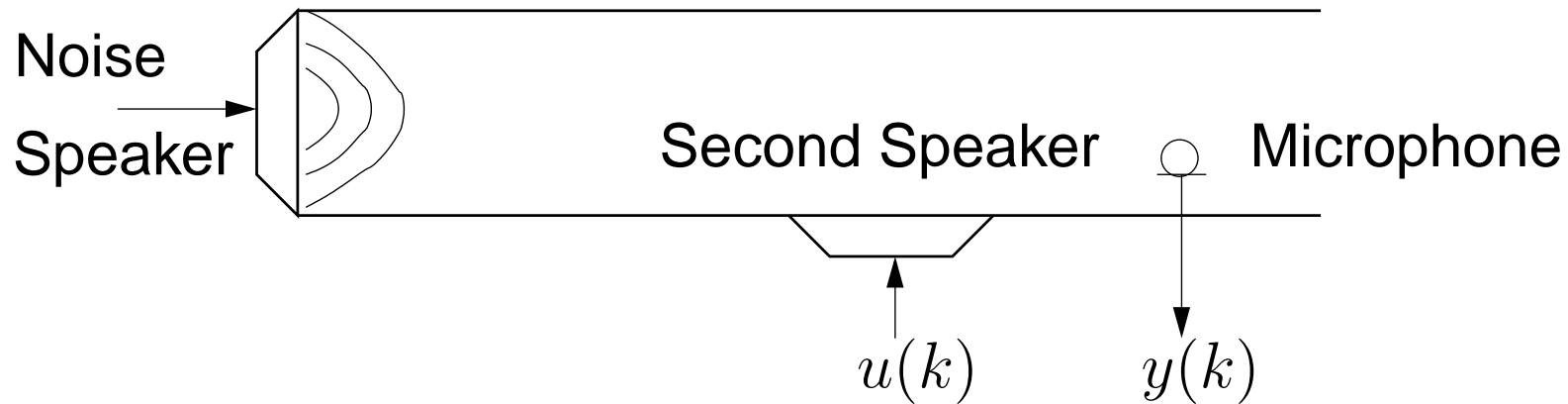
Challenges for the classical OE Identification approach

1. You need to fix the order of the SS model **a priori!**
2. The parametrization for MIMO systems is a **difficult?**
3. The optimization problem is **non-convex**, except for the FIR model.

linCost.m, nlCost.m

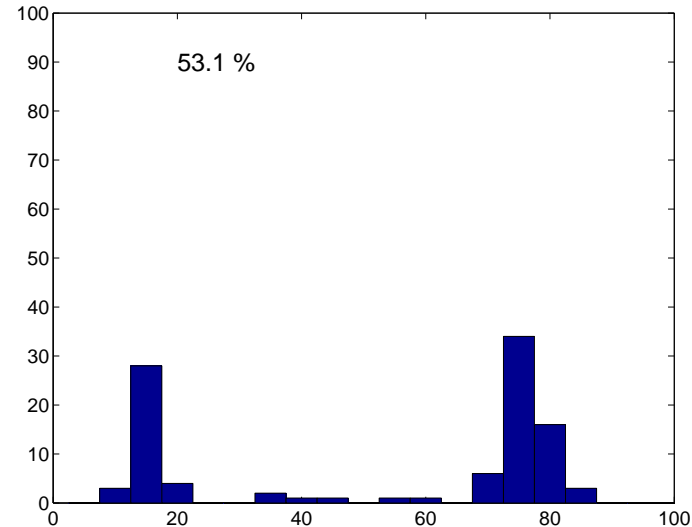
Consequence of non-convexity cost function

Example: OE model ($n=6$) for Acoustical duct



Results of a Monte Carlo simulation of 100 runs

Model error in percentage
100% = perfect fit



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The PE Identification recipe

1. Given the “true” **SGM**:

$$y(k) = G(q)u(k) + H(q)e(k) \in \mathbb{R} \quad E[u(k)e(\ell)^T] = 0 \text{ and} \\ e(k) \sim (0, R)$$

2. **Model $\widehat{\text{SGM}}(\theta)$** : many possibilities!

The AR(MA)X model:

$$y(k, \theta) = \frac{b_1 q^{-1} + \dots + b_n q^{-n}}{1 + a_1 q^{-1} + \dots + a_n q^{-n}} u(k) + \frac{1 + (c_1 q^{-1} + \dots + c_n q^{-n})}{1 + a_1 q^{-1} + \dots + a_n q^{-n}} e(k)$$

$$\text{with } \theta = [a_i, b_i, c_i, \dots].$$

Finding “optimal” predictor using SS innovation form

The AR(MA)X model:

$$y(k, \theta) = \frac{b_1 q^{-1} + \dots + b_n q^{-n}}{1 + a_1 q^{-1} + \dots + a_n q^{-n}} u(k) + \frac{1 + (c_1 q^{-1} + \dots + c_n q^{-n})}{1 + a_1 q^{-1} + \dots + a_n q^{-n}} e(k)$$

with $\theta = [a_i, b_i, c_i, \dots]$. This model parametrizes the state space innovation model as (Lemma 8.3):

$$x(k+1, \theta) = \underbrace{\begin{bmatrix} -a_1 & 1 & & 0 \\ & & \ddots & \\ & & & 1 \\ -a_{n-1} & & & \\ -a_n & & \dots & 0 \end{bmatrix}}_{A_n(a)} x(k, \theta) + \underbrace{\begin{bmatrix} b_1 \\ \vdots \\ b_{n-1} \\ a_n \end{bmatrix}}_{B_n(b)} u(k) + \underbrace{\begin{bmatrix} (c_1) - a_1 \\ \vdots \\ (c_n) - a_n \end{bmatrix}}_{K_n(k)} e(k)$$

$$y(k, \theta) = \underbrace{\begin{bmatrix} 1 & \dots & 0 \end{bmatrix}}_{C_n} x(k, \theta) + e(k)$$

What is the optimal predictor of the output?

The “optimal” predictor of an AR(MA)X model

The “optimal” predictor derived from the innovation form equals:

$$\hat{x}(k+1, \theta) = \begin{bmatrix} -c_1 & 1 & & 0 \\ & & \ddots & \\ & & & 1 \\ -c_{n-1} & & & \\ -c_n & \dots & & 0 \end{bmatrix} \hat{x}(k, \theta) + \begin{bmatrix} b_1 \\ \vdots \\ b_{n-1} \\ b_n \end{bmatrix} u(k) + \begin{bmatrix} k_1 \\ \vdots \\ k_{n-1} \\ k_n \end{bmatrix} y(k)$$

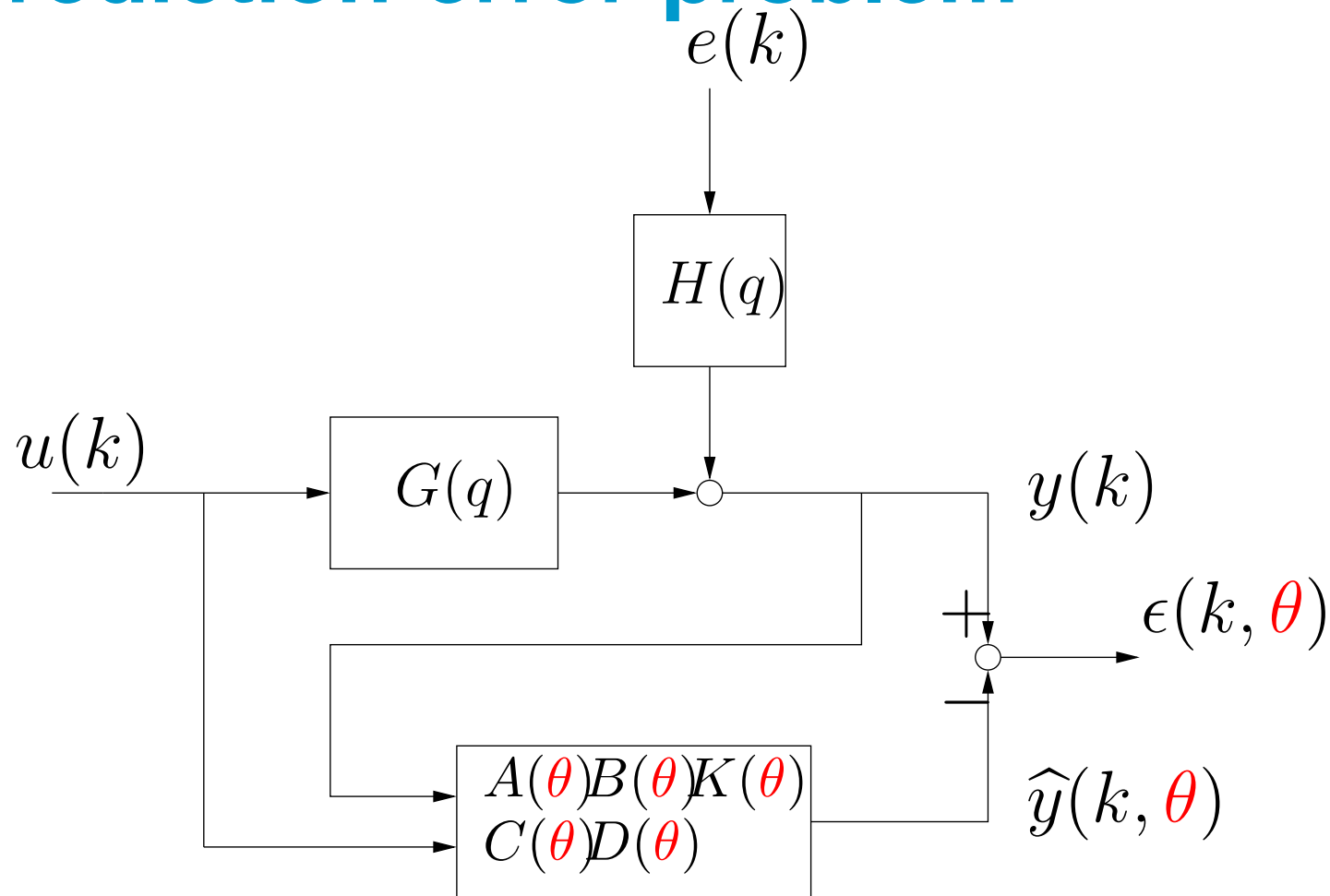
$$\hat{y}(k, \theta) = \begin{bmatrix} 1 & \dots & 0 \end{bmatrix} \hat{x}(k, \theta)$$

with $k_i = (c_i) - a_i$. This is given in transfer function form as:

$$\hat{y}(k, \theta) = \frac{B(q, \theta)}{C(q, \theta)} u(k) + \frac{\left(C(q, \theta) \right) - A(q, \theta)}{\left(C(q, \theta) \right)} y(k)$$

The ARX predictor is **linear** in the parameters a_i, b_i

Prediction-error problem

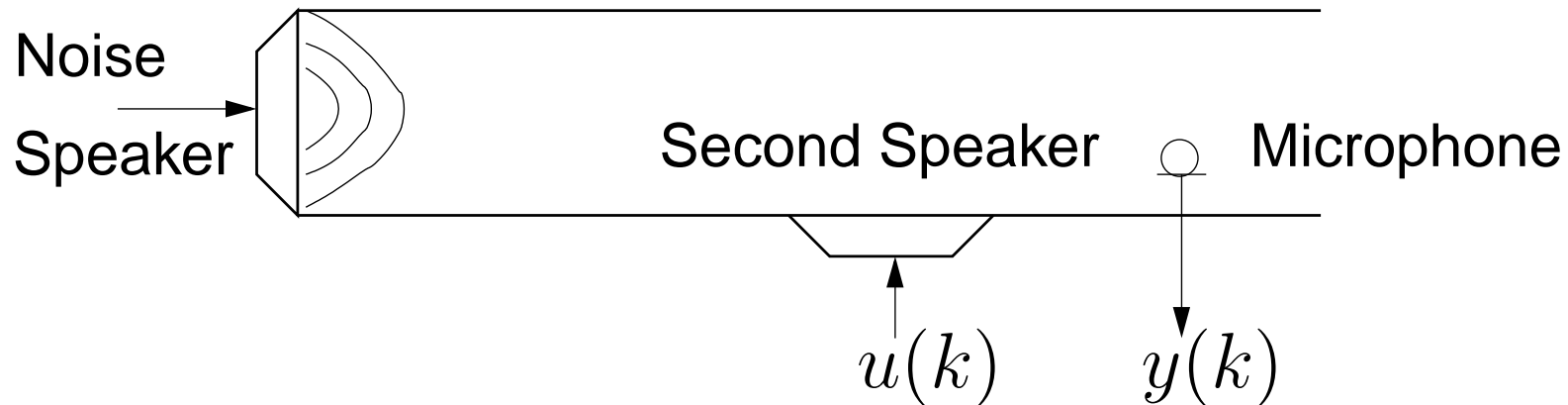


$$\hat{\theta}_N = \operatorname{argmin} J_N(\theta) \quad \text{with} \quad J_N(\theta) = \frac{1}{N} \sum_{k=1}^N \|y(k) - \hat{y}(k, \theta)\|_2^2$$

Challenges for the classical PEM Identification approach

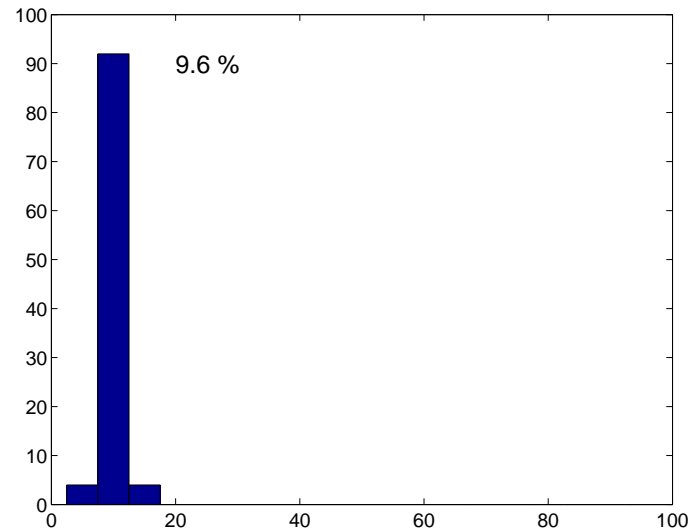
1. You need to fix the order of the SS model **a priori!**
2. The parametrization for MIMO systems is a **difficult?** and many possibilities to model the stochastic part!
3. The optimization problem is **non-convex**, except for the ARX model.

Example: ARX model (n=6) for Acoustical duct



Results of a Monte Carlo simulation of 100 runs

Model error in percentage
100% = perfect fit



Preparation for this afternoon

Preparation:

1. Study Chapter 5 (5.5.2 - 5.8) for the KF problem.
2. Study Chapter 8 (8.3) for the predictor parametrization.

Download Homework 3